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Power Law Type Long Memory Behaviors Modeled with Distributed Time Delay Systems

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Abstract: This paper studies a class of distributed time delay systems that exhibit power law type long memory behaviors. Such dynamical behaviors are present in multiple domains and it is therefore essential to have tools to model them. The literature is full of examples in which these behaviors are modeled by means of fractional models. However, several limitations of fractional models have recently been reported and other solutions must be found. In the literature, the analysis of distributed delay models and integro-differential equations in general is older than that of fractional models. In this paper, it is shown that particular delay distributions and conditions on the model coefficients make it possible to obtain power laws. The class of systems considered is then used to model the input-output behavior of a lithium-ion cell.

Keywords: power law behaviors; distributed time delay systems; fractional models; lithium-ion batteries

1. Introduction

Systems with long memory behaviors sometimes generate outputs whose power spectrum $X(\omega)$ exhibits power laws $(X(\omega) \approx K/\omega^{2\nu})$ when supplied by white noise. Numerous examples of this kind of behavior have been revealed in various domains:

- Electrochemistry through charge diffusion in batteries [1–4];
- Thermal conduction through the exact solution of the heat equation in a semi-infinite medium linking the heat rate to the surface temperature [5,6];
- Biology through complex dynamics in biological tissues [7];
- Mechanics through the dynamical properties of viscoelastic materials and in particular wave propagation problems in these materials [8];
- Acoustics through visco-thermal losses in wind instruments [9];
- Electrical distribution networks [10].

The fractional integrator operator that links an input u(t) to an output y(t) and defined in the Riemann–Liouville sense by the Equation [11]

$$y(t) = \int_0^t \frac{1}{(t-\tau)^{\nu-1}} u(\tau) d\tau = {}_0^{\nu} I\{u(t)\},\tag{1}$$

also exhibits this kind of behavior as the modulus of the Fourier transform of this operator, denoted $|Y(j\omega)/U(j\omega)|$ is defined by $1/\omega^{\nu}$. That is why a fractional order pseudo-state space description was introduced to model this kind of behavior. In the commensurate case and in most of the studies found in the literature, this description is defined by:

$$\begin{cases} \frac{d^{\nu}}{dt^{\nu}}x(t) = Ax(t) + Bu(t) \\ y(t) = Cx(t) \end{cases}$$
 (2)

where $\frac{d^{\nu}}{dt^{\nu}}$ is a fractional differentiation operator (which is, in a certain sense, the inverse operator to the operator ${}_{0}^{\nu}I\{$ }). There are several definitions for $\frac{d^{\nu}}{dt^{\nu}}$ in the literature [11]—more than 30 are listed in [12]—that are not all equivalent in relation to initial conditions [13–15].

Even if this kind of model provides a good fitting of the input-output behavior of many systems, Description (2) has several drawbacks:

- The variable x(t) that plays the role of the state does not have the properties of a state, which is why the name pseudo-state was introduced [13,14];
- The initial conditions are not well taken into account if the Caputo or the Riemann–Liouville definitions are used for the derivative of the pseudo-state; it is better to define the description with a fractional integration to take into account the model past [13–17];
- The use of fractional differentiation in the pseudo-state space description is not mandatory and only fractional integration is needed [16];
- This description memory is infinite and it exhibits infinitely slow and fast time constants (even if they are attenuated, they exist), which excludes the possibility of linking the pseudo-state variable to a physical variable [14,16];
- Exact observability cannot be reached as all the system past must be known to predict its future [18];
- The fractional integration given by Equation (1) involves a singular kernel [19,20]. As mentioned in [20], this leads to complications in the solution/simulation of the fractional order differential equations;
- In modeling, several mathematical and interpretation problems can invalidate the models obtained [21].

In recent decades, there has been a frantic "fractionalization" of concepts and methods dedicated to integer models. It therefore seemed necessary to better define the benefits and limits of fractional models. Over many years of research in the field of fractional models, I have largely participated in updating several of the drawbacks previously mentioned. Faced with the omnipresence of phenomena and systems that exhibit power law behaviors, I became convinced that it was necessary to better understand the physical origins of these phenomena and it is proposed here to use other ways of modeling them than fractional models.

Thus, to overcome these problems and to model power law type long memory behavior, a class of distributed time delay models will be used here. Distributed time delay models, also known in the literature as hereditary models [22] or integro-differential equation-based models [23], have long existed. In this study, the delay distribution used in their definition is particularized in order to obtain the desired power law behaviors. The class of model considered is then applied to the modeling of a lithium-ion cell input-output behavior.

2. A Class of Time Delay Systems That Exhibits a Power Law Long Memory Behavior

The following class of distributed time delay systems is considered.

$$\frac{d}{dt}x(t) = A_0x(t) + A_1 \int_0^{T_f} \eta(\tau)x(t-\tau)d\tau + Bu(t)$$
with $A_0 \in \mathbb{R}$, $A_1 \in \mathbb{R}$ and $B \in \mathbb{R}$ (3)

Distributed delay makes it possible to model the cumulative effect of the past values on the dynamics [3]. The literature abounds in practical applications of distributed delay systems, which appear in fields as varied as thermodynamics, ecology, epidemiology (predator–prey systems), logistics, traffic flow, microorganism growth, or hematopoiesis. In all these cases, the use of a distributed kernel allows thinner modeling of the interactions between the different system components.

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In Equation (3), kernel $\eta(t)$ is assumed to be defined by:

$$\eta(t) = C_0 \frac{t^{\nu - 1} e^{-\omega_l t}}{\Gamma(\nu)} \text{ with } 0 < \nu < 1\omega_l \in \mathbb{R}_+^*, \tag{4}$$

with

$$C_0 = (1 + \omega_l^2)^{\frac{\nu}{2}} \text{ with } \Gamma(\nu) = \int_0^\infty t^{\nu - 1} e^{-t} dt.$$
 (5)

The Laplace transform of Equation (3) (disregarding the initial conditions) is given by:

$$sx(s) = A_0x(s) + A_1 \int_0^{T_f} \eta(\tau)x(s)e^{-\tau s}d\tau + Bu(s)$$
 (6)

and thus

$$sx(s) = A_0x(s) + A_1x(s) \int_0^{T_f} \eta(\tau)e^{-\tau s}d\tau + Bu(s).$$
 (7)

The following integral is now studied:

$$I(s) = \int_0^{T_f} \eta(\tau) e^{-\tau s} d\tau. \tag{8}$$

Function $\eta(t)$ being given by Equation (4), integral I(s) is given by (see Appendix A)

$$I(s) = \frac{C_0 \left(T_f\right)^{\nu}}{\Gamma(1+\nu)^{1}} F_1 \left(\nu, 1+\nu, T_f(s+\omega_l)\right) \tag{9}$$

where ${}_{1}F_{1}(a,b,z)$ denotes the Kummer function defined by:

$${}_{1}F_{1}(a,b,z) = \sum_{n=0}^{\infty} \frac{(a)_{n}}{(b)_{n}} \frac{z^{n}}{n!}$$
(10)

with

$$(a)_n = a(a+1)(a+2)\dots(a+n-1), \quad (a)_0 = 1.$$
 (11)

Integral I(s) given by Equation (9) is represented in Figures 1 and 2 for several values of ν and for several values of T_f . These figures highlight that:

- Parameter ν affects the order of the power law behaviors;
- Parameter T_f chosen such that $T_f = 10/\omega_l$, controls the frequency band on which the power law behavior exists.

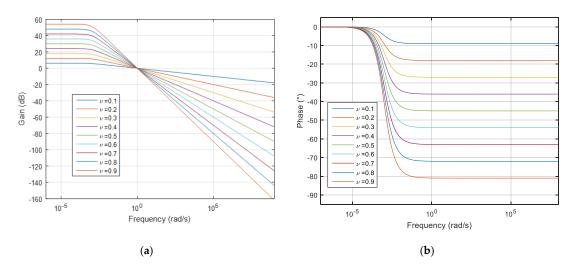


Figure 1. Gain (a) and phase (b) diagrams of integral I(s) for various values of ν and $T_f = 10,000$.

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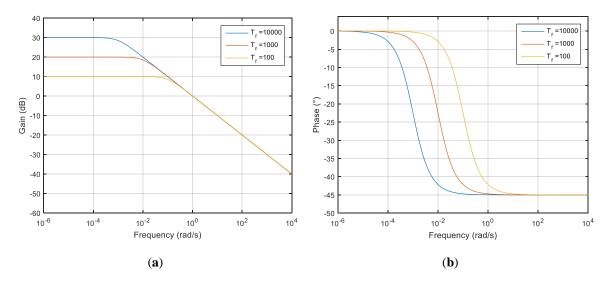


Figure 2. Gain (a) and phase (b) diagrams of integral I(s) for various values of T_f and $\nu = 0.5$.

Using expression I(s), Equation (7) becomes:

$$sx(s) = A_0x(s) + A_1x(s)I(s) + Bu(s).$$
 (12)

The transfer function linking x(s) to u(s) is given by:

$$\frac{x(s)}{u(s)} = \frac{B}{s - A_1 I(s) - A_0} \tag{13}$$

and can be rewritten as follows:

$$\frac{x(s)}{u(s)} = \frac{-B/A_1}{-s/A_1 + I(s) + A_0/A_1} = \frac{\beta}{\alpha_1 s + I(s) + \alpha_0} = H(s) \text{ with } \beta = \frac{-B}{A_1}, \ \alpha_1 = \frac{-1}{A_1}, \text{ and } \alpha_0 = \frac{A_0}{A_1}.$$
 (14)

Parameters β , α_0 , and α_1 can be used to control the gain of the transfer function H(s) (value of H(0)) and the frequency band on which the power law behavior takes place. Figure 3 describes how parameters (must) act on the gain asymptotic diagram of the H(s) denominator to obtain the required power law behavior. Let $\omega_1 = 10/T_f$. As $\left|I(j\omega)\right| \approx 1/\omega^{\nu}$ with $\omega \gg \omega_1$, in the configuration of Figure 4, it can be checked that:

$$\omega_2 \approx \sqrt{\left(\frac{C_0}{\alpha_0}\right)^{\frac{2}{\nu}} - \omega_1^2} \qquad \omega_3 \approx \frac{\alpha_0}{\alpha_1}.$$
 (15)

Figure 4 shows the frequency response H(s)/s (H(s) given by Equation (14)) for various values of ν and with $\omega_1=0.001$ rd/s, $\omega_2=10^4$ rd/s, $\omega_3=10^6$ rd/s, and $T_f=10^4$ s. This figure highlights the ability of the considered class of models to produce power law behaviors and shows how the parameters control the frequency band on which this behavior takes place.

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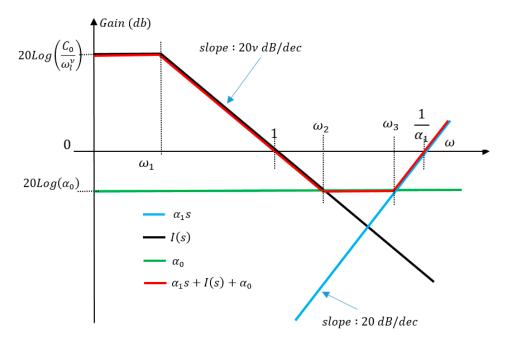


Figure 3. Impact of parameters ν , α_0 , and α_1 on the gain of Model (14) denominator.

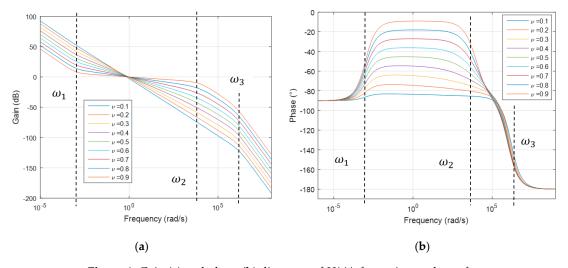


Figure 4. Gain (a) and phase (b) diagrams of H(s)/s for various values of ν .

3. Power Law Long Memory Behavior Without Singular Kernel

The kernel used in Equation (4) has a singularity at time t=0. To overcome such a problem, new operators were proposed in [23] and others studied in [19,20]. In this work, it is proposed to modify Equation (4) by the Equation:

$$\eta(t) = C_0 \left(\frac{\omega_l^{\nu}}{\Gamma(\nu)} t^{\nu-1} e^{-\omega_l t} - \frac{\omega_l^{\nu}}{\Gamma(\nu)} t^{\nu-1} e^{-\omega_m t} + \omega_l^{\nu} \omega_m^{1-\nu} e^{-\omega_m t} \right)$$
(16)

with

$$C_0 = \left| \left(\frac{1}{\left(\frac{j}{\omega_l} + 1 \right)^{\nu}} - \left(\frac{\omega_l}{\omega_m} \right)^{\nu} \frac{1}{\left(\frac{j}{\omega_m} + 1 \right)^{\nu}} + \left(\frac{\omega_l}{\omega_m} \right)^{\nu} \frac{1}{\frac{j}{\omega_m} + 1} \right)^{-1} \right|. \tag{17}$$

As t tends toward 0, the following relation holds (using Taylor expansions of the exponential function in Equation (18)):

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$$\eta(t) \approx C_0 \left(\frac{\omega_l^{\nu}}{\Gamma(\nu)} t^{\nu-1} (-\omega_l t + \omega_m t) + \omega_l^{\nu} \omega_m^{1-\nu} \right)$$
(18)

and thus

$$\lim_{t \to 0} \eta(t) = C_0 \omega_l^{\nu} \omega_m^{1-\nu}. \tag{19}$$

This highlights the non-singularity of Kernel given by Equation (14) as time t tends toward 0. With this new definition, the integral in Equation (8) is given by:

$$I(s) = K_1 \Big({}_1F_1 \Big(\nu, 1 + \nu, (s + \omega_l) T_f \Big) - {}_1F_1 \Big(\nu, 1 + \nu, (s + \omega_m) T_f \Big) \Big) + K_2 \Big(\frac{1 - e^{-(s + \omega_m) T_f}}{s + \omega_m} \Big)$$
(20)

with

$$K_{1} = \frac{C_{0}\omega_{l}^{\nu}T_{f}^{\nu}}{\Gamma(1+\nu)} \qquad K_{2} = C_{0}\omega_{l}^{\nu}\omega_{m}^{1-\nu}.$$
 (21)

Figure 5 shows the gain and phase diagrams of integral I(s) for various values of ν and with $\omega_l = 10^{-2}$ rd/s and $\omega_m = 10^4$ rd/s.

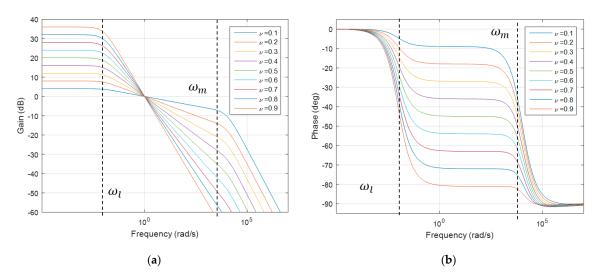


Figure 5. Frequency response (gain on the (a), phase on the (b)) of I(s) with the non-singular kernel $\eta(t)$ for ν varying from 0.1 to 0.9 and with $\omega_l = 10^{-2}$ rd/s, $\omega_m = 10^4$ rd/s.

As for the kernel given by Equation (4), the parameters A_0 , A_1 , and B in Model (14), and thus the parameters of β , α_0 , and α_1 , can be used to control the frequency band on which the power law behavior takes place. This is illustrated by Figure 6 that shows the impact of α_0 on the frequency response of H(s).

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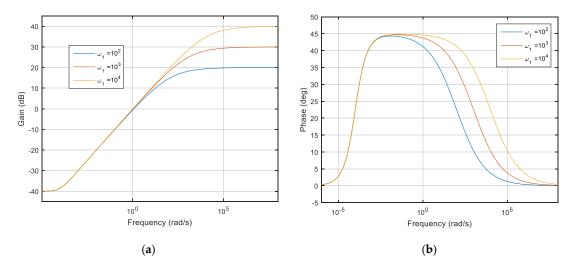


Figure 6. Frequency response H(s) for various values of $\alpha_0 = 1/\omega_1^{\nu}$, with $\nu = 0.5$: gain (a) and phase (b).

4. Application

It is proposed in this section to use the class of models introduced above to capture the behavior of a lithium-ion cell. The need to implement short-term memory models such as the one introduced in this paper for the modeling of electrochemical devices was highlighted in [24]. Also, it was shown in [25] that only a positive electrode can be considered and under the hypothesis of a single spherical particle for this electrode, a lithium-ion cell can be modeled using the block diagram of Figure 7, in which:

- A system S_d models the diffusion of lithium in the spherical particle and links the current I(t) to the concentration of lithium $C_s(t)$ at the surface of the spherical particle;
- A nonlinear function $OCV(t) = f(C_S(t))$ links the concentration of lithium at the surface of the spherical particle $C_S(t)$ to the open circuit voltage OCV(t);
- A resistor *R* is used to model the cell internal resistance and contact resistance.

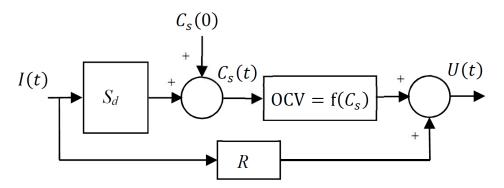


Figure 7. Model of a lithium-ion cell proposed in [25].

The function $OCV(t) = f(C_s)$ can be obtained by measuring the cell voltage at rest for various cell states of charge (SOC). For the considered cell, this function is shown in Figure 8 and fitted by a degree 10 polynomial. The initial state of the battery is defined by $C_S(0)$ and the cell voltage is U(t).

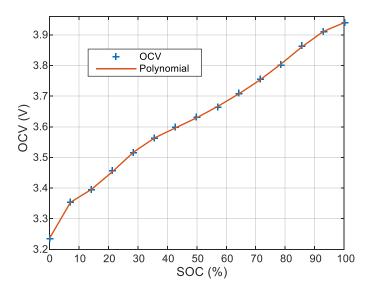


Figure 8. Considered cell open circuit voltage (OCV) fitted by a degree 10 polynomial.

The diffusion part exhibits a power law behavior at high frequency and it is proposed here to model it using the following model:

$$\begin{cases} \frac{d}{dt}x(t) = A_1 \int_0^{T_f} \eta(\tau)x(t-\tau)d\tau + BI(t) \\ U(t) = A_1 \int_0^t x(\tau)d\tau \end{cases}$$
 (22)

in which the kernel $\eta(\tau)$ is defined by Equation (16) with $\omega_l = 10/T_f$ and ω_m chosen ten times greater than the sampling frequency of the current and voltage signals.

The current profile of Figure 9 is applied to the cell, and the parameters of Model (22) are tuned using a nonlinear optimization algorithm that aims to reduce the error between the measured voltage and the model output in Figure 7.

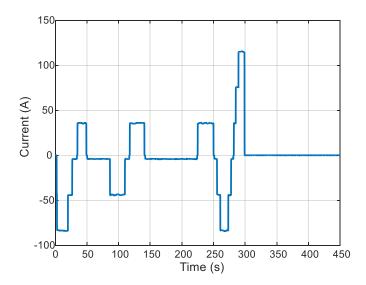


Figure 9. Current profile applied to the cell.

The parameter optimization process provides the following results for the Model:

$$A_1 = 989.6, B = 45.92, T_f = 1410 s$$
 (23)

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Figure 10 shows a comparison of the cell voltage and the voltage provided by the model. It reveals a good accuracy of the model and confirms the ability of this class of model to capture long memory power law type behaviors. The differences that appear are the result of the very high currents applied to the cell that exhibit some non-linear behaviors not taken into account by the model used.

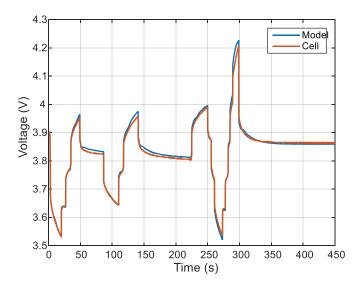


Figure 10. Comparison of the cell voltage and the voltage provided by the model.

5. What Does the Proposed Approach Solve?

The previous sections show that the class of distributed delay models considered (Equation (3)) is able to exhibit a power law behavior on a controlled frequency band. In comparison with the fractional Model (2), and in relation with the comments in the introduction section:

- In Equation (3), the variable x(t) can be viewed as a real state and a physical meaning can be associated to it;
- There is no longer any ambiguity in the operator used for the definition of Equation (3) (in Equation (2), the Caputo, Riemann–Liouville, or other operators [12] can be chosen);
- The memory of Model (3) is of finite length;
- Initialization the Model (3) requires the knowledge of its state on a finite length and is well defined.

6. Conclusions

Dynamic systems that exhibit power law behaviors are currently modeled in the literature using fractional models. Several papers recently revealed various limitations of this class of model. In order to overcome these limitations, this paper is a first attempt to show that power law behaviors can also be captured by classes of models that were introduced into the literature some time ago, i.e., distributed delay models [26,27] (but also the Volterra equation of the second kind [28,29]). Two delay weight functions permitting the desired power law are proposed. First, a simple one, but with a singularity, and a second one without. For these two functions, conditions on the model coefficients are given for the definition of the frequency band on which the power law takes place. The class of model considered is used to capture the input-output behavior of a lithium-ion cell, a system that is known to exhibit a power law behavior.

The author intends now to study in greater detail the properties of this class of model in order to reach more complex behaviors. Particular attention will be paid to the meaning that can be given to the state in a modeling approach. Other classes of model will also be used.

Conflicts of Interest: The authors declare no conflict of interest.

Appendix A

This appendix gives some indications on how to obtain Equation (9). In [30], the following result can be found:

$$\int_{0}^{1} e^{zt} t^{a-1} (1-t)^{b-a-1} dt = \frac{\Gamma(b-a)\Gamma(a)}{\Gamma(b)} {}_{1}F_{1}(a,b,z) \text{ with } \Re(b) > \Re(a) > 0$$
 (A1)

where $_1F_1(a,b,z)$ denotes the Kummer function defined by the Equation:

$$_{1}F_{1}(a,b,z) = \sum_{n=0}^{\infty} \frac{(a)_{n}}{(b)_{n}} \frac{z^{n}}{n!} \text{ with } (a)_{n} = a(a+1)(a+2)\dots (a+n-1), (a)_{0} = 1.$$
 (A2)

Note that the Kummer function (or confluent hypergeometric function of the first kind) is implemented in MATLAB software as kummer U(a, b, z).

The Equation (8) is given by:

$$I(s) = \int_0^{T_f} \eta(\tau) e^{-\tau s} d\tau = \int_0^{T_f} C_0 \frac{\tau^{\nu - 1} e^{-\omega_I \tau}}{\Gamma(\nu)} e^{-\tau s} d\tau.$$
 (A3)

Using the change of variable $x = \frac{\tau}{T_f}$ and thus $dx = \frac{1}{T_f} d\tau$, Integral (A3) can be rewritten as:

$$I(s) = \frac{C_0(T_f)^{\nu}}{\Gamma(\nu)} \int_0^1 x^{\nu - 1} e^{-xT_f(s + \omega_l)} dx.$$
 (A4)

Using a = v, b = 1 + v and $z = T_f(s + \omega_l)$ in Equation (A1), Integral (A4) is also defined by:

$$I(s) = \frac{C_0(T_f)^{\nu}}{\Gamma(1+\nu)} {}_1F_1(\nu, 1+\nu, T_f(s+\omega_l)). \tag{A5}$$

A similar calculation can be done with the kernel in Equation (16).

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