

Article A Nonlinear Subspace Predictive Control Approach Based on Locally Weighted Projection Regression

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Abstract: Subspace predictive control (SPC) is a widely recognized data-driven methodology known for its reliability and convenience. However, effectively applying SPC to complex industrial process systems remains a challenging endeavor. To address this, this paper introduces a nonlinear subspace predictive control approach based on locally weighted projection regression (NSPC-LWPR). By projecting the input space into localized regions, constructing precise local models, and aggregating them through weighted summation, this approach handles the nonlinearity effectively. Additionally, it dynamically adjusts the control strategy based on online process data and model parameters, while eliminating the need for offline process data storage, greatly enhancing the adaptability and efficiency of the approach. The parameter determination criteria and theoretical analysis encompassing feasibility and stability assessments provide a robust foundation for the proposed approach. To illustrate its efficacy and feasibility, the proposed approach is applied to a continuous stirred tank heater (CSTH) benchmark system. Comparative results highlight its superiority over SPC and adaptive subspace predictive control (ASPC) methods, evident in enhanced tracking precision and predictive accuracy. Overall, the proposed NSPC-LWPR approach presents a promising solution for nonlinear control challenges in industrial process systems.

Keywords: data-driven control; industrial process; locally weighted projection regression (LWPR); nonlinear system control; subspace predictive control (SPC)

1. Introduction

Industrial processes constitute the backbone of modern economies, contributing to diverse sectors such as chemical engineering, transportation, and energy production [1–3]. The efficient operation and regulation of these processes are essential for achieving optimal resource utilization, product quality, and safety [4]. In pursuit of these objectives, the field of industrial process control has emerged as a crucial discipline, aiming to harness advancements in science and technology to enhance process performance, stability, and reliability [5].

The significance of industrial process control extends beyond mere operational efficiency. It plays a pivotal role in ensuring consistent product quality, minimizing waste, and mitigating environmental impact [6–8]. Furthermore, effective control strategies empower industries to adapt swiftly to changing market demands and regulatory requirements, fostering competitiveness and sustainability [9]. However, the realm of industrial process control is not without its challenges. Conventional model-based control approaches encounter limitations when applied to complex industrial systems [10–12]. A prominent constraint is the difficulty in obtaining accurate and comprehensive model information. Constructing a detailed mathematical representation for complex processes is often formidable, especially given nonlinear dynamics, intricate interactions, and inherent uncertainties [13]. These challenges hinder the efficacy of conventional model-based control, leading to suboptimal performance, compromised stability, and difficulties in real-time adaptation [14].



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Copyright: © 2024 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https:// creativecommons.org/licenses/by/ 4.0/). To tackle these formidable challenges, data-driven control approaches have emerged as a promising and dynamic solution in the realm of industrial process control [15]. These methodologies, including machine learning [16], deep learning [17], and reinforcement learning [18], leverage the wealth of information derived from sensors, actuators, and historical data to formulate effective and adaptive control strategies.

Within the spectrum of data-driven control strategies, the subspace predictive control (SPC) approach stands out as a particularly compelling choice. SPC ingeniously combines subspace identification techniques with predictive control methodologies, rendering it an attractive option for its simplicity and ease of implementation [19]. The adoption of SPC has spurred extensive research into its practical applications within the realm of industrial process control. For instance, Li et al. [20] have devised a SPC method to regulate the power allocation of server racks and control the supply temperature of cold air. Furthermore, Navalkar et al. [21] have introduced a repetitive SPC approach that demonstrates precise individual blade pitch control on a wind turbine prototype. These applications underscore the potential of SPC in optimizing complex industrial systems. Nevertheless, it remains clear that many intricate industrial systems inherently exhibit nonlinear behaviors, intricate interactions, and uncertain dynamics. While the strength of SPC lies in its foundation on linear models and its reliance on offline data, it encounters formidable challenges when confronted with the inherent complexities of nonlinear processes. This limitation has the potential to curtail its effectiveness in capturing the multifaceted nature of these systems and responding adeptly to dynamic variations [22].

To address the prevalent issue of the inadequacy of the SPC method in dealing with the intricate nonlinear dynamics inherent in industrial processes, there is a compelling imperative to develop and advance the field of adaptive subspace predictive control (ASPC). The primary objective motivating ASPC is to facilitate real-time adjustments of controller parameters in response to dynamic data fluctuations, offering a dynamic and adaptive approach to control. A cornerstone of ASPC involves the utilization of a sliding data window mechanism, which serves as a vital tool for describing the current operational conditions and effectively mitigating the nonlinear complexities often encountered in intricate systems. This approach has been notably applied and refined by pioneering researchers such as Wahab et al. [23], Vajpayee et al. [24], and Hallouzi et al. [25]. Their work has showcased the effectiveness of the sliding data window in applications ranging from wastewater treatment systems to nuclear reactors and even complex models like the Boeing 747 aircraft. While substantial progress has been achieved in the application of ASPC, a noteworthy limitation lies in the fact that these methodologies have predominantly been tailored to linear controllers. This limitation restricts their capability to comprehensively address the intricate nonlinear characteristics commonly found in diverse industrial scenarios.

To achieve a more robust and appropriate solution, some researchers have explored alternative avenues by directly crafting controllers explicitly designed for specific nonlinear systems. For example, a specialized nonlinear subspace predictive controller tailored to bilinear systems is introduced in [26]. Zhou et al. [27] and Luo et al. [28] have extended nonlinear SPC methods to encompass Hammerstein systems and Hammerstein–Wiener systems, expanding the scope of applicability. However, it is essential to recognize that while these endeavors have shown promise, designing controllers for specific nonlinear systems often lacks the necessary universality required for broad industrial implementation. In light of these considerations, the field of nonlinear SPC is confronted with the challenge of achieving a more versatile solution.

In this paper, a nonlinear subspace predictive control approach based on locally weighted projection regression (NSPC-LWPR) is presented to address the aforementioned issues. The locally weighted projection regression (LWPR) algorithm, which is an incremental nonparametric statistical learning technique [29] and is related to the field of linear parameter varying modeling [30–32], is integrated into the SPC method. By fitting the local nonlinear relationships between input and output data to construct a predictive model, higher prediction

accuracy can be achieved when the expected output of the nonlinear process changes, while maintaining smooth tracking. The main contributions of this approach are listed as follows:

- (1) Seamless integration of LWPR and SPC: The LWPR algorithm and the SPC method are seamlessly integrated for industrial process control. By projecting the input space into localized regions, constructing precise local models, and aggregating them through weighted summation, the proposed approach effectively addresses the complex nonlinear relationships in industrial processes.
- (2) Enhanced adaptability and efficiency: The proposed approach constructs the controller from the trained regression model. This implies that it can adapt the control strategy using online process data and local model parameters. In addition, it removes the necessity for storing offline process data. These advancements highlight improvements in both adaptability and efficiency.
- (3) Improved predictive and tracking performance: The proposed approach shows improvements in both predictive and tracking performance. It creates an accurate predictive model by capturing the dynamic characteristics of the system from input/output (I/O) data. This boosts the accuracy of the predictive controller, especially during transitions from nonlinear to steady-state processes. The increased prediction accuracy also greatly enhances the tracking performance of the predictive controller. In situations where the expected output of the nonlinear process changes, the controlled system adjusts smoothly to match the projected output path, ensuring consistent and smooth tracking.

This paper is structured as follows. Section 2 offers an extensive elucidation of the preliminaries associated with the subspace predictor and the LWPR learning scheme. Section 3 focuses on the design of the controller, including parameter determination criteria and theoretical analysis. The application of the proposed NSPC-LWPR approach in a CSTH benchmark study is showcased in Section 4. Finally, Section 5 concludes the paper by summarizing its main content and suggesting potential directions for future research.

2. Preliminaries

2.1. Subspace Predictor

Assuming discrete time intervals indexed by *k* where measurements of the I/O data for the system are denoted by $u_k \in \mathbb{R}^m$ and $y_k \in \mathbb{R}^l$, the stacked vector $u_{s,k}$ of length *s* is introduced as

$$u_{s,k} = \begin{bmatrix} u_k^T & \cdots & u_{k+s-1}^T \end{bmatrix}^T.$$
(1)

The block Hankel matrices U_p and U_f are constructed as

$$U_{p} = \begin{bmatrix} u_{s_{p},k-s_{p}+1} & \cdots & u_{s_{p},k-s_{p}+\bar{N}} \end{bmatrix},$$

$$U_{f} = \begin{bmatrix} u_{s_{f},k+1} & \cdots & u_{s_{f},k+\bar{N}} \end{bmatrix},$$
(2)

where the indexes p and f correspond to the past and future block Hankel matrices, respectively. s_p and s_f both denote the number of row blocks. \bar{N} represents the sample length. Similarly, the output data block Hankel matrices Y_p and Y_f are defined based on the output data.

The subspace predictor model represents the optimal prediction of Y_f as a combination of past I/O data and future input data [33]. The subspace predictor can be formulated as

$$\hat{Y}_f = L_w W_p + L_u U_f, \tag{3}$$

where L_w and L_u are the subspace predictor coefficient matrices, and $W_p = \left[\left(Y_p \right)^T \left(U_p \right)^T \right]^T$.

2.2. LWPR Learning Scheme

The LWPR algorithm employs the standard regression model $y = \beta^T x + \varepsilon$ to approximate the nonlinear function $y = f(x) + \varepsilon$, where *x* is the input vector, *y* is the scalar output, and ε is a zero-mean random noise term.

To capture the locality aspect, the position of each data point *x* is leveraged through a Gaussian kernel to compute the weight *w*:

$$w = \exp\left(-0.5(x - x_c)^{\mathrm{T}} D(x - x_c)\right), \quad 0 < w \le 1,$$
(4)

where x_c denotes the center of a local subset of data, and D is a positive semi-definite distance metric that determines both the size and shape of the neighborhood contributing to the establishment of the corresponding local model. A smaller D results in a smoother kernel, while a larger D captures finer details. As discussed in [34], besides the Gaussian kernel, alternative kernel functions can also be employed. However, the choice of kernel function only affects the computation of weights and consequently influences the number and shape of local models, but it does not significantly impact the prediction results.

Based on the obtained weights, the following weighted means can be calculated:

$$\bar{x} = \sum_{n=1}^{N} w_n x_n \Big/ \sum_{n=1}^{N} w_n, \bar{y} = \sum_{n=1}^{N} w_n y_n \Big/ \sum_{n=1}^{N} w_n.$$
(5)

By subtracting \bar{x} and \bar{y} from the original measurements, the input and output of the LWPR algorithm can be guaranteed with zero means.

Following the initialization of LWPR without a locally linear model (receptive field, RF), the algorithm proceeds with the training process. For each training sample, the weight is computed using (4). Subsequently, the regressions, projections, and distance metrics of each RF are updated iteratively until no new RF creation is required. The crucial aspects of the LWPR learning scheme for one RF centered at x_c , which hold relevance for our extension of locally weighted learning to SPC, are concisely summarized in Table 1. Corresponding symbols and their notations are provided in Table 2.

Table 1. Locally Weighted Projection Regression (LWPR) learning scheme for one RF centered at x_c [29].

```
1. Initialization: (number of training samples seen n = 0)
       x_0^0 = 0, \beta_0^0 = 0, W^0 = 0, u_r^0 = 0, p_r^0 = 0; r = 1 : R
2. Incorporating new data: Given training point(x,y)
       2a. Compute activation and update the means
               1.w = \exp(-0.5(x-x_c)^T D(x-x_c)); W^{n+1} = \lambda W^n + w
              2.x_0^{n+1} = (\lambda W^n x_0^n + wx) / W^{n+1};
                 \beta_0^{n+1} = \left(\lambda W^n \beta_0^n + wy\right) / W^{n+1}
       2b. Compute the current prediction error
              x_{res,1} = x - x_0^{n+1}, \hat{y} = \beta_0^{n+1}
               Repeat for r = 1: R(projections)
                1.z_r = x_{\text{res},r}^T u_r^n / \sqrt{u_r^n u_r^n} u_r^n
                2.\hat{y} = \hat{y} + \beta_r^n z_r
                3.x_{\operatorname{res},r+1} = x_{\operatorname{res},r} - z_r p_r^n
                 4.\mathrm{MSE}_r^{n+1} = \lambda \mathrm{MSE}_r^n + w(y - \hat{y})^2
       2c. Update the local model
       res_1 = y - \beta_0^{n+1}
       Repeat for r = 1 : R(projections)
              2c.1 Update the local regression and compute residuals
                1.a_{zz,r}^{n+1} = \lambda a_{zz,r}^n + wz_r^2; a_{zres,r}^{n+1} = \lambda a_{zres,r}^n + wz_r res_r2.\beta_r^{n+1} = a_{zres,r}^{n+1} / a_{zz,r}^{n+1}
                3.\operatorname{res}_{r+1} = \operatorname{res}_r - z_r \beta_r^{n+1}
                4.a_{\mathrm{xz},r}^{n+1} = \lambda a_{\mathrm{xz},r}^n + w x_{\mathrm{res},r} z_r
               2c.2 Update the projection directions
                 1.u_r^{n+1} = \lambda u_r^n + w x_{\text{res},r} \text{res}_r
                 2.p_r^{n+1} = a_{xz,r}^{n+1} / a_{zz,r}^{n+1}
```

Notation	Description
Ν	Number of training data points
М	Number of local models
R	Number of local projections
$z_r(r=1:A)$	<i>r</i> th element of the lower-dimensional projection of input data <i>x</i>
$\mathbf{u}_r(r=1:A)$	<i>r</i> th projection direction
$\mathbf{p}_r(r=1:A)$	Regressed input space to be subtracted to maintain orthogonality of projection directions
W	Diagonal weight matrix representing the activation due to all samples
$\beta_r(r=1:A)$	<i>r</i> th component of slope of the local linear model $\beta = \begin{bmatrix} \beta_1 & \cdots & \beta_R \end{bmatrix}^T$
λ	Forgetting factor used to exclude data and accelerate the learning process
MSE_r^n	Mean square error of the <i>n</i> th sample in the <i>r</i> th projection
$a_{\text{zz},r}^n, a_{\text{zres},r}^n, a_{\text{xz},r}^n$	Sufficient statistics for incremental computation of r th dimension of variable var after seeing n data points

Table 2. Indexes and symbols used for LWPR [29].

3. Locally Weighted Projection Regression-Based Subspace Predictive Control

3.1. Controller Design

Considering that only the leftmost column of \hat{Y}_f is considered to predict the output, (3) can be rewritten as

$$\hat{y}_{N_p l} = \hat{L}_w w_p + \hat{L}_u u_{N_c m},\tag{6}$$

where \hat{y}_{N_pl} is the first N_pl row of the leftmost column in \hat{Y}_f , w_p is the leftmost column of W_p , and u_{N_cm} is the first N_cm row of the leftmost column in u_f . \tilde{L}_w and \tilde{L}_u are truncated from L_w and L_u .

Given the congruity in structure between the subspace predictor outlined in (3) and the regression model employed for approximating nonlinear functions within the framework of the LWPR algorithm, it follows that the LWPR algorithm becomes instrumental in the computation of the coefficients \tilde{L}_w and \tilde{L}_u for the subspace predictor. Then, for the query point u_{N_cm} , the calculation of the *i*-th element of its output vector $\hat{y}_{N_pl}^i$ in the local prediction output $\hat{y}_{N_nl}^{i,j}$ of the *j*-th locally linear model can be simplified as follows:

$$\hat{y}_{N_{p}l}^{i,j} = \beta_0^i + \sum_{r=1}^R \beta_r^{i,j} s_r^{i,j},$$
(7)

where $1 \le i \le N_p l$, $1 \le j \le M$, β_0^i is the average of the *i*-th training output samples calculated in (5). $\beta_r^{i,j}$ signifies the parameter linked to the respective RF, while $s_r^{i,j}$ is defined as

$$s_{1}^{i,j} = (u_{1}^{i,j})^{T} \vartheta,$$

$$s_{2}^{i,j} = (u_{2}^{i,j})^{T} \left(I - p_{1}^{i,j} (u_{1}^{i,j})^{T}\right) \vartheta,$$

$$\vdots$$

$$s_{R}^{i,j} = (u_{R}^{i,j})^{T} \left(\prod_{r=R-1}^{1} \left(I - p_{r}^{i,j} (u_{r}^{i,j})^{T}\right)\right) \vartheta,$$
(8)

where $\vartheta = u_{N_cm} - \tilde{u}_{N_cm}$, and \tilde{u}_{N_cm} is the average of the training input samples.

Then, $\hat{y}_{N_n l}^{i,j}$ can be rewritten as

$$\hat{y}_{N_{pl}}^{i,j} = \zeta^{i,j} + L^{i,j} u_{N_{c}m},\tag{9}$$

where

$$\zeta^{i,j} = \beta_0^i - \sum_{r=1}^R \psi_\tau \tilde{u}_{N_c m}, L^{i,j} = \sum_{r=1}^R \psi_r,$$
(10)

$$\psi_{r} = \begin{cases} \beta_{1}^{i,j}(u_{1}^{i,j})^{T}, r = 1\\ \beta_{r}^{i,j}(u_{r}^{i,j})^{T} \left(\prod_{d=r-1}^{1} \left(I - p_{d}^{i,j}\left(u_{d}^{i,j}\right)^{T}\right)\right). r \neq 1 \end{cases}$$
(11)

Based on the obtained weights, the total output $\hat{y}_{N_pl}^i$ of the LWPR model is the normalized weighted mean of all the predicted outputs $\hat{y}_{N_pl}^{i,j}$ of the *M* local models, that is,

$$\hat{y}_{N_{p}l}^{i} = \sum_{j=1}^{M} \omega_{j} \hat{y}_{N_{p}l}^{i,j} / \sum_{j=1}^{M} \omega_{j}.$$
(12)

To better understand the solving process of global output $\hat{y}_{N_{pl}}^{i}$, the information processing unit of the LWPR learning scheme is shown in Figure 1.



Figure 1. Information processing unit of the LWPR learning scheme.

Furthermore, we have

$$\hat{y}_{N_{pl}}^{i} = L_{cst}^{i} + L_{cft}^{i} u_{N_{c}m}, \tag{13}$$

where

$$L_{cst}^{i} = \sum_{j=1}^{M} \omega_{j} \zeta^{i,j} / \sum_{j=1}^{M} \omega_{j}, L_{cft}^{i} = \sum_{j=1}^{M} \omega_{j} L^{i,j} / \sum_{j=1}^{M} \omega_{j}.$$
(14)

Then, \hat{y}_{N_pl} can be expressed as

$$\hat{y}_{N_p l} = L_{cst} + L_{cft} u_{N_c m} \tag{15}$$

where

$$L_{cst} = \begin{bmatrix} \left(L_{cst}^{1} \right)^{T} & \cdots & \left(L_{cst}^{i} \right)^{T} & \cdots & \left(L_{cst}^{N_{pl}} \right)^{T} \end{bmatrix}^{T}, L_{cft} = \begin{bmatrix} \left(L_{cft}^{1} \right)^{T} & \cdots & \left(L_{cft}^{i} \right)^{T} & \cdots & \left(L_{cft}^{N_{pl}} \right)^{T} \end{bmatrix}^{T}.$$
(16)

To enhance the precision of the system's behavior modeling and maintain the consistent accuracy of predictions, it is advisable to express the projected output in (13) through an incremental formulation concerning Δu_{N_cm} :

$$\hat{y}_{N_{p}l} = A_{N_{p}l}^{1} y_{k} + A_{N_{p}l}^{2} L_{cft} \Delta u_{N_{c}m},$$
(17)

where

$$A_{N_{p}l}^{1} = \begin{bmatrix} I_{l} \\ I_{l} \\ \vdots \\ I_{l} \end{bmatrix}, A_{N_{p}l}^{2} = \begin{bmatrix} I_{l} & 0 & \cdots & 0 \\ I_{l} & I_{l} & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ I_{l} & I_{l} & \cdots & I_{l} \end{bmatrix},$$

$$\Delta u_{N_{c}m} = \begin{bmatrix} (\Delta u_{k+1})^{T} (\Delta u_{k+2})^{T} \cdots (\Delta u_{k+N_{c}m})^{T} \end{bmatrix}^{T},$$
(18)

while I_l is the identity matrix in the dimension of l, and Δu_{k+1} in Δu_{N_cm} is defined as

$$\Delta u_{k+1} = u_{k+1} - u_k, \tag{19}$$

and other components, such as Δu_{k+2} and Δu_{k+N_cm} , are defined similarly as Δu_{k+1} .

The approach is designed to generate a control signal u_k by minimizing a quadratic cost function *J*. This cost function takes into account the incremental input Δu_k , the provided reference signal r_k , and the projected output \hat{y}_k , and is mathematically expressed as follows:

$$J = \sum_{n_i=1}^{N_p} (\tilde{Q})^T W_Q(\tilde{Q}) + \sum_{n_j=1}^{N_c} (\tilde{R})^T W_R(\tilde{R}),$$
(20)

where

$$\tilde{Q} = r_{k+n_i} - \hat{y}_{k+n_i}, \quad \tilde{R} = \Delta u_{k+n_i}, \tag{21}$$

while N_p and N_c are the prediction and control horizons. W_Q and W_R are the weighting matrices of the cost function *J*.

Based on (20) and (17), J can be rewritten as

$$J = (\bar{Q})^{T} W_{Q}(\bar{Q}) + (\bar{R})^{T} W_{R}(\bar{R}),$$
(22)

where \bar{Q} and \bar{R} are represented as

$$\bar{Q} = A_{N_p l}^1 (r_k - y_k) + A_{N_p l}^2 \tilde{L}_u \Delta u_{N_c m}, \bar{R} = \Delta u_{N_c m}.$$
(23)

Based on (22), it becomes evident that the cost function is exclusively contingent upon Δu_{N_cm} , a quantity attainable through the computation of the derivative of the cost function with respect to Δu_{N_cm} under unconstrained circumstances (UCs). Furthermore, the differential quandary can be redefined as a quadratic problem under constrained circumstances (CCs). Consequently, the representation of Δu_{N_cm} takes the form:

$$\Delta u_{N_cm} = \begin{cases} \frac{\partial J}{\partial \Delta u_{N_cm}} = 0, & UC\\ \min_{\Delta u_{N_cm}} J \ s.t.A_{QP} \le B_{Qp}, & CC \end{cases}$$
(24)

where A_{QP} and B_{QP} are constructed from preset constraints.

Upon acquiring Δu_{N_cm} , the initial *m* components are chosen for utilization. Drawing from (19) and armed with the understanding of u_k as well as Δu_{k+1} , one can ascertain the forthcoming controller output to be incorporated into the regulated system, thereby determining u_{k+1} .

Subsequently, the control diagram outlining the proposed NSPC-LWPR approach is depicted in Figure 2, while Algorithm 1 succinctly encapsulates the essential steps.

Notably, the training of this locally weighted regression model is closely related to the number of inputs and outputs of the MIMO system. Specifically, the number of inputs and outputs directly impacts the training cost and computational efficiency of the algorithm. In essence, a larger number of inputs and outputs in the MIMO system increases the



model's training cost, reduces the computational efficiency, and prolongs the runtime. Conversely, a small number will have the opposite effect.

Figure 2. The control diagram of the proposed nonlinear subspace predictive control approach based on locally weighted projection regression (NSPC-LWPR) approach.

Algorithm 1 The proposed NSPC-LWPR approach.

Step 1. Initialization

- a. Fully excite the initialization signal of the system input;
- b. Initialize the LWPR model with no RF;
- Step 2. LWPR Regression Model Training
- a. Normalize the process data;
- b. Train the LWPR regression model utilizing the learning scheme specified in II-B;

c. Continue the training until the predicted output of the controlled object consistently converges to its actual value;

Step 3. Subspace Predictor Construction

- a. Calculate the subspace predictor's coefficients according to (16);
- b. Convert the predictor to the incremental form of (17);

Step 4. Control Input Signal Calculation

- a. Select methods to find Δu_{N_cm} based on (24);
- b. Calculate the control input according to (19);
- c. Denormalize the solved control signal and input it into the controlled system;

Step 5. Judgment

a. If the controlled system is still running, return to **Step 2** and calculate the next signal; b. If the controlled system stops running, the proposed approach is terminated and the calculation of the next control signal is stopped.

Furthermore, to emphasize the superiority of the proposed NSPC-LWPR approach, a theoretical comparison is performed between it and the MPC, SPC, and ASPC methods as delineated in Table 3.

Table 3. Theoretical comparison among different control strategies.

Method	MPC	SPC	ASPC	NSPC-LWPR
Approach Type	model-based	data-driven	data-driven	data-driven
Prior Knowledge	model information	off-line process data	no need	no need
Dynamic Ability	able	unable	able	able
Controller Type	fixed; linear	fixed; linear	unfixed; linear	unfixed; nonlinear

3.2. Parameters Determination Criteria

Achieving a balance between the computational efficiency and effectiveness of the proposed control approach relies heavily on making careful choices regarding parameters such as s_p , s_f , N_c , N_p , W_Q , and W_R . These parameter selections are critical in ensuring that the control system operates smoothly and effectively. The specific details are as follows.

3.2.1. s_p and s_f

 s_p and s_f correspond to the number of row blocks contained within the past and future Hankel matrices. Choosing an excessively large value for these parameters can result in a model that has too many parameters, potentially causing problems related to complexity. On the other hand, selecting a value that is too small may result in a model with too few parameters, potentially affecting its accuracy and predictive abilities.

3.2.2. N_c and N_p

The choice of the control horizon N_c influences the behavior of the control signal and the control law's structure, while the predictive horizon N_p is crucial for tracking error calculations. It is recommended to set Nc to be greater than or equal to the system's order α for precise control performance, and N_p should be larger than N_c to ensure effective tracking, within the limits defined by the predictive horizon s_f . Care must be taken to strike a balance, as selecting excessively large N_c and N_p values can increase computational demands, particularly in fast systems, while overly small values may compromise effectiveness. In total, the criteria for determining N_c and N_p are $\alpha + 1 \le N_c \le N_p \le s_f$, with N_c shaping the control signal and N_p affecting the tracking accuracy.

3.2.3. W_O and W_R

 W_Q and W_R are employed as adjustable parameters in the optimization process, serving to impose penalties on the tracking error and the rate of control signal variation, respectively. Opting for substantial penalties on tracking errors yields a swifter yet potentially more aggressive response, facilitating rapid adaptation. Conversely, assigning a substantial penalty to the control signal engenders a more resilient but potentially slower controller, fostering stability and reducing abrupt changes in control action.

3.3. Theoretical Analysis

For the convenience of the theoretical analysis, the cost function in (20) can be rewritten as

$$J = \sum_{n_i=1}^{N_p} \left(\breve{Q} \right)^T W_Q \left(\breve{Q} \right) + \sum_{n_j=1}^{N_c} \left(\breve{R} \right)^T W_R \breve{R}, \tag{25}$$

where

$$Q = h_{k+n_i|k+1} = r_{k+n_i|k+1} - \hat{y}_{k+n_i|k+1},$$

$$\tilde{R} = g_{k+n_i|k+1} = \Delta u_{k+n_i|k+1}.$$
(26)

while $k + n_i | k + 1$ represents the prediction at the $(k + n_i)$ th sampling time when the current time is k + 1.

Then, the sequences of $\phi_{h,k+1}$ and $\phi_{g,k+1}$ are given by

$$\phi_{h,k+1} = \left\{ h_{k+2|k+1}, h_{k+3|k+1}, \dots, h_{k+N_p+1|k+1} \right\},
\phi_{g,k+1} = \left\{ g_{k+1|k+1}, g_{k+2|k+1}, \dots, g_{k+N_c|k+1} \right\}.$$
(27)

Based on the descriptions mentioned above, the dynamic of the controlled system can be modeled with the following nonlinear discrete-time difference equations:

$$h_{k+2|k+2} = f\left(h_{k+1|k+1}, g_{k+1|k+1}\right),\tag{28}$$

and the problem to be solved at step k + 1 can be turned into

Problem*:
$$\min_{\phi_{g,k+1}} J(h_{k+1})$$

 $s.t.: g_{k+n_j|k+1} \in G, n_j \in \{1, ..., N_c\};$
 $h_{k+n_i|k+1} \in H, n_i \in \{2, ..., N_p\};$
 $h_{k+1+N_n|k+1} \in H_t, H_t \in H,$
(29)

where *G* is the time-invariant set, *H* is the convex constraints set subject to the system evolution, and the terminal set H_t is $\{0\}$.

Theorem 1. *The proposed control approach is recursively feasible, and the controlled system under the proposed control approach is asymptotically stabilized at the origin.*

Proof of Theorem 1. The sequence $\phi_{g,k+1}^*$, which is assumed to be the optimal solution to Problem* at step k + 1, is represented as

$$\phi_{g,k+1}^* = \left\{ g_{k+1|k+1}^*, g_{k+2|k+1}^*, \dots, g_{k+N_c|k+1}^* \right\},\tag{30}$$

and the corresponding optimal sequence $\phi_{h,k+1}^*$ is given by

$$\phi_{h,k+1}^* = \left\{ h_{k+2|k+1}^*, h_{k+3|k+1}^*, \dots, h_{k+N_p+1|k+1}^* \right\}.$$
(31)

Since $h_{k+N_p+1|k+1}^* \in H_t$ according to (29) applies, and H_t equals to {0}, we can obtain that $\Phi(\Psi) \in G$, and

$$h_{k+N_p+2|k+1} = f(\Psi, \Phi(\Psi)) \in H_t, \tag{32}$$

where $\Psi = h_{k+N_p+1|k+1}^*$. The terminal controller Φ exists such that $\Phi(x) \in G$ for all $x \in H_t$, and $f(x, \Phi(x)) \in H_t$ for all $x \in H_t$ under the condition that H_t , which equals to $\{0\}$, is a control invariant set of the system. $\Phi(\Psi)$ characterizes the effect of the terminal controller Φ on Ψ .

The temporary sequences $\phi_{g,k+2}^{tp}$ and $\phi_{h,k+2}^{tp}$ are given by

$$\phi_{g,k+2}^{tp} = \left\{ g_{k+2|k+1}^{*}, \cdots, g_{k+N_{c}+1|k+1}^{*}, \Phi(\Psi) \right\},
\phi_{h,k+2}^{tp} = \left\{ h_{k+3|k+1}^{*}, \cdots, h_{k+N_{p}+1|k+1}^{*}, h_{k+N_{p}+2|k+1} \right\},$$
(33)

where $\phi_{g,k+2}^{tp}$ and $\phi_{h,k+2}^{tp}$ both satisfy constraints of Problem*, and $\phi_{g,k+2}^{tp}$ is a feasible solution of the proposed approach to the Problem* after moving to $h_{k+2|k+1}^*$ at step k + 2.

Based on the analysis provided above, if a feasible solution to Problem* exists for k = 1, it implies that there is also a feasible solution for the problem at any $k \in \{1, 2, 3, \dots\}$. Therefore, it can be concluded that the proposed control approach, developed by solving Problem*, is recursively feasible.

In what follows, the stability analysis of the proposed control approach is presented. The difference in cost between $J^{cd}(h_{k+2})$ and $J^*(h_{k+1})$ can be computed from

$$J^{cd}(h_{k+2}) - J^{*}(h_{k+1}) = \left(h_{k+N_{p}+2|k+1}\right)^{T} W_{Q}\left(h_{k+N_{p}+2|k+1}\right) - \left(h_{k+2|k+1}^{*}\right)^{T} W_{Q}\left(h_{k+2|k+1}^{*}\right) - \left(g_{k+1|k+1}^{*}\right)^{T} W_{R}\left(g_{k+1|k+1}^{*}\right) + \left(\Phi\left(h_{k+N_{p}+1|k+1}^{*}\right)\right)^{T} W_{R}\left(\Phi\left(h_{k+N_{p}+1|k+1}^{*}\right)\right),$$
(34)

where cost $J^{cd}(h_{k+2})$ is led by the sequence $\phi_{g,k+2}^{tp}$ and $\phi_{h,k+2}^{tp}$ at step k + 2, and $J^*(h_{k+1})$ is the optimal cost at step k + 1.

Since both $h_{k+N_p+2|k+1}$ and $h^*_{k+N_p+1|k+1}$ belong to H_t , it is evident that the right-hand side of (34) is nonpositive. Additionally, $J^{cd}(h_{k+2})$ serves as an upper bound for the optimal cost $J^*(h_{k+2})$. Therefore, we can derive the following result:

$$J^*(h_{k+2}) \le J^{cd}(h_{k+2}) \le J^*(h_{k+1}).$$
 (35)

Based on the fact that J^* decreases monotonically as the Lyapunov function, it can be concluded that the controlled system, governed by the solution to Problem^{*}, satisfies $J^*(h_{k+2}) \leq J^*(h_{k+1})$.

Consequently, the controlled system is asymptotically stabilized at the origin. This completes the proof. $\hfill\square$

4. Benchmark Study on Continuous Stirred Tank Heater

The continuous stirred tank heater (CSTH) is a vital component in various industrial processes, particularly in the field of chemical engineering. This reactor is designed for the purpose of simultaneously heating and mixing fluid substances. It comprises tanks equipped with both mixing and heating elements, allowing for a continuous flow of fluids in and out of these tanks, thereby ensuring constant movement. During this process, the fluids are subjected to heating through various methods such as electric heaters or steam injection. Concurrently, sophisticated mixing mechanisms are employed to maintain uniform temperatures and prevent the formation of temperature gradients within the system. Precise control over essential variables, including temperature and flow rates, is crucial to optimizing heat transfer efficiency and facilitating desired reaction kinetics. In this paper, the CSTH system has become a valuable platform for evaluating the effectiveness of the proposed NSPC-LWPR approach.

As shown in Figure 3, the Automation Laboratory within the Department of Chemical Engineering at IIT Bombay has developed a widely acknowledged CSTH system [35]. It comprises five distinct inputs and three resultant outputs. Specifically, inputs u_1 , u_2 , and u_3 correspond to flow rates that are governed by individual valves, while inputs u_4 and u_5 pertain to the intensity of heating within two distinct heaters. The three outputs of the system encompass the temperature of the first tank T_1 , the temperature of the second tank T_2 , and the water level within the second tank h_2 .



Figure 3. The continuous stirred tank heater (CSTH) system in IIT Bombay.

Considering the needs during the production process, u_4 and u_5 are considered the two adjustable input variables, and T_1 is the predetermined setpoints. The remaining parameters are set to their steady-state values shown in Table 4 unless otherwise specified.

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Parameter	Description	Value
V1	Volume of tank 1	$1.75 \times 10^{-3} \text{ m}^3$
A_2	Cross sectional area of tank 2	$7.584 imes 10^{-3} \text{ m}^2$
r_2	Radius of tank 2	0.05 m
U	Heat transfer coefficient	235.1 W/m ² K
T_c	Cooling water temperature	30 °C
T_a	Atmospheric temperature	25 °C
u_1	Flow F_1 (%Input)	60%
<i>u</i> ₂	Flow F_2 (%Input)	55%
u_3	Flow F_R (% Input)	50%
u_4	Heat input Q_1 (% Input)	60%
u_5	Heat input Q_2 (% Input)	80%
T_1	Steady state temperature (tank 1)	49.77 °C
T_2	Steady state temperature (tank 2)	52.92 °C
h_2	Steady state level	0.3599 m

Table 4. Nominal model parameters and steady state.

To further enhance the tracking control performance, the smoothing approximation, which is a filtering process, is introduced to make the expected output able to change smoothly from one desired state to the other. Specifically, the expected temperature for T_1 , denoted as $y_{sp}^{T_1}$, is set to be

$$y_{sp}^{T_{1}}(k) = \begin{cases} y_{sp1}^{T_{1}}, k \in (0, 600] \\ \lambda_{T_{1}} y_{sp}^{T_{1}}(k-1) + (1-\lambda_{T_{1}}) y_{sp2}^{T_{1}}, k \in (600, 1300] \\ \lambda_{T_{1}} y_{sp}^{T_{1}}(k-1) + (1-\lambda_{T_{1}}) y_{sp1}^{T_{1}}, k \in (1300, 2000] \end{cases}$$
(36)

in which $y_{sp1}^{T_1} = 50$, $y_{sp2}^{T_1} = 52$, and the smoothing coefficient denoted as λ_{T_1} is set to be 0.998. To account for the mechanical constraints of the CSTH system, the predictive controller is subject to the constraints with $30 \le y \le 60$, $0 \le u \le 100$, and $-0.5 \le \Delta u \le 0.5$.

The parameters setup of the CSTH benchmark study is illustrated in Table 5, where E_Q and E_R are the eigenvalues of W_Q and W_R , and f_s is the sampling frequency.

Table 5. Parameters setup of the CSTH benchmark study.	
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Parameter	E_Q	E_R	s_p	s_f	N_c	N_p	f_s	
Value	1	2	10	5	3	4	10 Hz	

The outputs and setpoints of T_1 under various control frameworks are presented in Figure 4. It is evident that the outputs of T_1 exhibit inadequate setpoint tracking performance under the SPC framework. The output curves show erratic behavior, characterized by shaking and oscillations, making the tracking of setpoints ineffective during set point changes. Conversely, the tracking performance under the ASPC framework improves, effectively following setpoints after a settling time. However, during the setpoint of T_1 transitions, the tracking performance under the proposed NSPC-LWPR framework shows the most promising results among the three control approaches. Even when the setpoints of T_1 change at 600 s and 1300 s, the T_1 outputs consistently track the setpoints.

The disparities observed in the tracking performance of T_1 can be attributed to variations in the subspace predictor outputs generated by the controllers under different control frameworks as illustrated in Figure 5. The SPC method employs a fixed, offline-designed subspace predictor, making it unsuitable for effectively controlling nonlinear systems. In contrast, the ASPC method incorporates online learning capabilities to optimize its parameters based on process data, enabling it to adapt to changing conditions and generate corresponding subspace predictor outputs dynamically. However, the ASPC method remains linear and approximates new conditions using fixed nearby sampling points. Consequently, this approach can lead to degraded tracking performance and overshooting issues during smooth setpoint changes. Similar to the ASPC method, the proposed NSPC-LWPR approach is equipped with autonomous learning capabilities, allowing real-time updates of controller information using newly generated process data. However, it surpasses the limitations of the ASPC controller by employing multiple linear working points for weighted summation. This innovative approach constructs a nonlinear subspace predictive controller, leading to a more precise predictive output for the current operating condition.







Figure 5. Subspace predictor outputs u_4 and u_5 under different control frameworks. (**a**) Under the SPC framework. (**b**) Under the ASPC framework. (**c**) Under the NSPC-LWPR framework.

The results indicate that the proposed NSPC-LWPR approach excels in describing the current nonlinear operating condition and achieves superior tracking performance in nonlinear industrial process control compared to SPC and ASPC methods. This highlights its potential as an advanced and effective controller in nonlinear industrial process system applications.

According to Figures 4 and 5, the outputs of the controlled system and subspace predictor are strictly limited within specific boundaries. Additionally, the change rates of u_4 and u_5 are investigated and shown in Figure 6. It can be observed that the values of Δu_4 and Δu_5 both fall within the range of -0.5 and 0.5, which aligns with the constraints set. This result indicates that the proposed NSPC-LWPR approach operates within the imposed constraints, which effectively influence the system behavior.



Figure 6. Change rates of u_4 and u_5 .

To demonstrate the predictive performance of the proposed NSPC-LWPR approach, we conduct a comparative analysis using the multi-step prediction means squared error (MPMSE), defined as:

$$\sigma(N_p) = \sum_{k=T_s}^{T_t} \sum_{j=1}^{N_p} \|\hat{y}(k+j) - y(k+j)\|^2 / (N \cdot N_p),$$
(37)

where T_s and T_t represent the starting point and the terminal point of the sampling data considered for analysis. To account for the necessary initialization time required by the proposed NSPC-LWPR approach, we set T_s to be 400, and T_t to be 10,000. The multi-step prediction mean squared error comparison among different control algorithms with varying predictive horizon N_p is presented in Table 6.

Control Methods	$N_p = 3$	$N_p = 4$	$N_p = 5$	$N_p = 6$	$N_p = 7$
SPC	1.5429	1.2649	1.0923	0.9732	0.8983
ASPC	0.1321	0.1147	0.1026	0.0934	0.0896
NSPC-LWPR	0.0946	0.0845	0.0740	0.0711	0.0661

Table 6. Multi-step prediction mean squared error (MPMSE) comparison.

Table 6 reveals a clear trend in the MPMSE, where the proposed NSPC-LWPR approach consistently outperforms the ASPC method and significantly outpaces the SPC method, all while maintaining a constant value of N_p . This performance discrepancy can be attributed to the SPC method's limited ability to effectively control nonlinear systems, leading to subpar output predictions. In contrast, both the ASPC and NSPC-LWPR approaches exhibit self-learning capabilities, enhancing their control of nonlinear systems. However, the proposed NSPC-LWPR approach stands out by demonstrating superior predictive accuracy under the current operational conditions.

Furthermore, as N_p increases, MPMSE decreases across all three methods. This decline is attributed to the broader prediction range, resulting in higher prediction accuracy. The outcomes of this comparative analysis compellingly support the superiority of the proposed NSPC-LWPR approach in terms of predictive performance. This finding indirectly substantiates its efficacy in enhancing tracking capabilities, underscoring its potential for effective control in nonlinear industrial process systems. In total, data analysis in this study was conducted using Python programming language (V3.9.2) with the following libraries: NumPy (V1.21.6), SciPy (V1.11.0), and Matplotlib (V3.3.1) [36].

5. Conclusions

In this paper, we propose a NSPC-LWPR approach to address tracking issues in nonlinear industrial process control. Our approach integrates the LWPR algorithm into the framework of the SPC method, harnessing the exceptional nonlinear handling capabilities of LWPR. Through the segmentation of the input space into localized regions, the construction of precise local models, and their aggregation through weighted summation, our approach adeptly captures dynamic system characteristics and trains the regression model. The adaptability and efficiency of our approach are further augmented by a dynamic control strategy that adjusts based on the online process data and the parameters of established local models. Furthermore, the verification of our approach against the CSTH benchmark unequivocally demonstrates its superiority over conventional SPC and ASPC methods. This verification affirms its ability to significantly enhance tracking precision and predictive accuracy in industrial process control.

While our proposed control approach has demonstrated exceptional performance, it is important to acknowledge that there remain unexplored avenues for further research. Future investigations could delve into methodological refinements, expanding the applicability of our approach to diverse control problems, or exploring advanced variants of the LWPR algorithm.

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